



0 009 291
A1

EUROPEAN PATENT APPLICATION

⑤ Int. Cl.³: **G 06 K 15/02, B 41 J 19/30**

②② Date of filing: 18.09.79

(71) Applicant: Svenska Philipsföretagen AB, Patent,
 S-11584 Stockholm (SE)
 (84) Designated Contracting States: SE

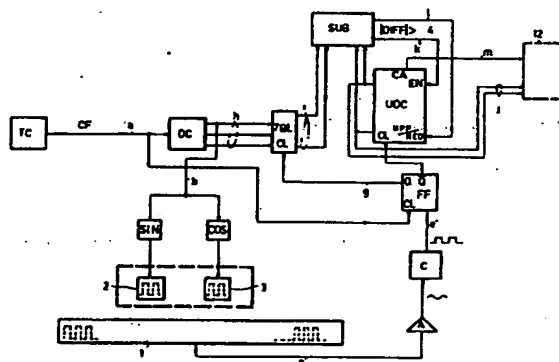
(71) Applicant: N.V. Philips' Gloeilampenfabrieken,
 Emmasingel 29, NL-5611 AZ Eindhoven (NL)
 (84) Designated Contracting States: DE FR GB IT

(72) Inventor: Thern, Rolf Ake Birger, c/o INT. OCTROOIBUREAU B.V. 6, Prof. Holstiaan, NL-5656 AA Eindhoven (NL)
Inventor: Dahlberg, Peter, c/o INT. OCTROOIBUREAU B.V. 6, Prof. Holstiaan, NL-5656 AA Eindhoven (NL)

74 Representative: Cuppens, Hubertus Martinus Maria et al, INTERNATIONAAL OCTROOIBUREAU B.V. Prof. Holstlaan 6. NL-5656 AA Eindhoven (NL)

(54) Method and device for controlling the carriage movement in a printer.

(57) In a printer comprising a printing head which is supported by a carriage (5) the carriage position along a record carrier is sensed at two occasions. A belonging microprocessor (12) calculates an actual carriage speed on basis of the position values which are obtained. The actual carriage speed and an optimal carriage speed which is stored, are compared and on basis of this comparison a compensation factor is calculated for a reference signal (Ref) which determines the carriage speed. This compensation factor is checked at each carriage movement by measuring the carriage speed, and thereby a permanently optimal carriage speed is secured and the need of an accurate trimming of the carriage speed is eliminated at the manufacture as well as at later service actions.



ACTORUM AG

BEST AVAILABLE COPY

Express Mail Label No.
EV332063212US

EP 0 009 291 A1

31.8.1979

1

PHZ 78/012C

Method and device for controlling the carriage movement in a printer.

The invention is directed to a method and a device for controlling the carriage movement in a printer comprising a printing head supported by the carriage and which is moved along a record carrier by the carriage movement, in which said carriage is moved by means of a direct current motor supported by the movable carriage or, alternatively, being stationarily mounted in the printer, the direct current motor being supplied by a drive unit which compensates for appearing variations of load during the carriage movement. The invention is also directed to a device for carrying out the method.

In printers of the type in question direct current motors are ordinarily used for driving the carriage thereof. Then the direct current motor will have its supply voltage from a drive unit being normally arranged to compensate for appearing load variations during the carriage movement by detection of the current which is supplied to the direct current motor. As a consequence of differences between different motor units trimming of the drive unit is performed at the manufacture of the printer, said drive unit being thereby adjusted so as to generate a supply voltage to the actual direct current motor of such a value that a given desirable and optimal carriage speed is obtained. As a consequence of the fact that some long-time deviation in the drive unit will always appear, as also mechanical wear of used mechanical parts for guiding the carriage, and the change of load of the motor which follows therefrom, said trimming operation will have to be renewed at later service actions.

The object of the invention is to provide a method for controlling the carriage movements as disclosed in the introduction of this description and which eliminates the need of trimming at the manufacture as well as later

31.8.1979

2

PHZ 78/012C

service actions.

The object of the invention is obtained by means of a method which according to the invention is characterized in that the drive unit is supplied with a reference signal via a digital-to-analogue converter from a microprocessor comprised in the printer, the carriage being thereby moved along at a carriage speed which corresponds to said reference signal, in that the carriage position along the record carrier is detected at two occasions when the carriage has reached the constant carriage speed which corresponds to said reference signal, in that the position information, after being eventually digitalized, is supplied to said microprocessor for calculating the actual carriage speed, in that the so calculated carriage speed is compared with an optimal carriage speed value, which stored in the microprocessor, and a compensation factor for the actual reference signal is calculated on basis of the comparison, in that a reference signal, which has been modified by said compensation factor, is supplied to the drive unit during the next following carriage movement, and in that said compensation factor is checked during each carriage movement by repeating the indicated steps. In the method according to the invention is used a so-called open loop control for controlling said motor. The advantage of a control loop of this type in this connection is that appearing oscillations and appearing tolerances of the mechanical transmission which is used for driving the carriage will not initiate self oscillations as a consequence of a feed back phase deviation in the control loop, which may follow from the use of a closed loop control. Start and turning of the carriage will not either give rise to any problems when using the method according to the invention since the speed measurement, with a knowledge of the driving motor characteristic, is not performed until the carriage has certainly obtained a constant speed.

According to the invention a device for controlling the carriage movement in a printer comprising a printing head which is supported by the carriage and which

31.8.1979

3

PHZ 78/012C

is moved along a record carrier by the carriage movement, a direct current motor for moving said carriage and being supported by the movable carriage or, alternatively, being stationarily mounted in the printer, and a drive unit for supplying a drive signal to said direct current motor, in which the drive unit compensates for appearing variations of load during the carriage movement, is characterized in that it is also provided with a carriage position indicator and a timing circuit for detecting the position of the carriage in relation to a position reference point at two occasions determined by said timing circuit, a signal processing circuit for analogue-to-digital conversion of the carriage position values which are obtained and for supplying said values to a microprocessor, said microprocessor being arranged to calculate therefrom an actual carriage speed between said two occasions and to compare this calculated speed with a stored, optimal carriage speed, and to calculate, on basis of said comparison, a compensation factor for a reference signal value which is generated by the microprocessor and which is supplied to said drive unit via a digital-to-analogue converter during the next following carriage movement.

One embodiment of the device according to the invention will be described closer in the following with reference to the drawings, in which

Fig. 1 shows schematically the components of the device according to the invention;

Fig. 2 shows a circuit diagram of the signal processing circuit of Fig. 1; and

Fig. 3 shows a signal diagram of signals appearing at different points in Fig. 2.

For detecting the carriage position this embodiment uses a so-called Meander-indicator comprising a stationary Meander-element 1 arranged in the printer and being extended in the direction of movement of the carriage along the printing width of the printer. On said carriage two further Meander-elements 2, 3 are arranged, the elements 2 and 3 being arranged so as to

31.8.1979

4

PHZ 78/012C

obtain a physical displacement of the belonging Meander-structures which is equal to $1/4$ period of the Meander-structure (corresponding to an electrical phase deviation of 90°) with respect to the Meander-structure provided on said stationary indicator element 1. The distance between the indicator element 1 and the indicator element 2, 3 is so selected that a good inductive coupling is obtained between the stationary element and the movable elements. To the movable elements 2, 3 are supplied, by the signal processing circuit 4, two sinus signals having a relative phase deviation of 90° . The signal, which is then induced in the stationary indicator elements 1, is supplied to said signal processing circuit 4. The operation and structure thereof will be described closer in connection with Fig. 2.

Fig. 1 discloses furthermore a schematical carriage 5 and a belonging drive motor 6, which via toothed belt 7 drives a gear wheel 8 which is engaged with a toothed belt 9, being fixed to the printer. When motor 6 is rotated the carriage is moved by means of the gear wheel 8 and the toothed belt 9.

The direct current motor is supplied by a supply voltage from a drive circuit 10, said circuit being supplied with a reference signal Ref from a digital-to-analogue converter 11, which in turn is supplied with a digital reference signal value from a microprocessor 12. Further is evident from the drawing that said microprocessor 12 is provided with a crystal controlled timing circuit TC which may be used for settling the occasions when the position of the carriage should be detected.

The used drive circuit 10 is of a type which is commonly used in tape recorders and record players. As shown within block 10 of Fig. 1 the drive circuit comprises a differential amplifier to the +input of which the Ref signal is supplied and the output of which is connected to the one pole of motor 6 and also feed-back coupled to the -input of the amplifier via a second resistor. For current sense the second pole of motor 6 is connected to earth via a third resistor and also to the +input of the amplifier

31.8.1979

5

PHZ 78/012C

via a fourth resistor.

In Fig. 2 the signal processing circuit 4 of Fig. 1 is disclosed and the indicator elements 1, 2, 3 which are connected thereto and also the crystal timing circuit TC being accommodated within the microprocessor 12 in Fig. 1.

The basic frequency of the microprocessor is 6.144 MHz and this frequency is divided by the factor of 2 so as to obtain from the timing circuit the frequency of 3.07 MHz, said frequency being designated as clock frequency CF in the following.

The clock frequency CF is supplied to a frequency divider DC comprising a counter counting down the clock frequency by the factor of 128 and thereby a signal of the approximate frequency of 24 kHz is obtained. This frequency is designated as primary frequency in the following.

Said primary frequency is supplied to a sinus current generator SIN and a cosinus current generator COS. The output signals from said generators are each supplied to a separate one of said Meander-elements 2 and 3, respectively, being designated as primary elements in the following. The primary elements are provided on a common circuit board which in turn is supported by the carriage 5.

For the period of the Meander-structure is selected the value of 3.92 mm and this period divided by 128 (approximately 0.0302 mm) is the smallest detectable change of the carriage position and is the subdistance which in a printer comprising a matrix print head, is used for obtaining the distance between two columns in the character matrix which is used for printing of characters having different character distances, e.g. 1/15, 1/12 or 1/10 inch.

The signals of the primary elements 2, 3 are supplied inductively to (the secondary) Meander-element 1. The secondary element 1 comprises a Meander-structure of the same period as elements 2, 3.

In element 1 the induced signals are added and thereby a signal of a constant amplitude is formed having a phase value in relation to the primary frequency

31.8.1979

6

PHZ 78/012C

signal which varies proportionally to the movement of the primary elements 2, 3 along the secondary element 1. The phase deviation will pass by the value of 0 (a multiple of 360°) each time when the Meander-structures of elements 1, 2, 3 reach a defined relative position, i.e. each time after 3.92 mm.

If the output signal from generators SIN and COS are $E \sin \omega t$ and $E \cos \omega t$, respectively, the output signal e of the secondary element 1 will be as follows

$$e = E \sin \omega t \cos \beta + E \cos \omega t \sin \beta = E \sin(\omega t + \beta)$$

in which β is the value of the phase angle which depends on the movement.

The voltage e is amplified in the amplifier A and is pulse shaped in the converter C to a rectangular wave form the frequency (the secondary frequency) of which will consequently correspond to the primary frequency (approximately 24 kHz) supplied by the frequency divider DC, and having a phase deviation (β) with respect to the primary frequency which depends on the carriage position. Now remains the conversion of this phase deviation to a digital value to be supplied to the microprocessor 12. In order to obtain the position of the carriage also the zero crossings of the phase deviation must be recorded.

For this object is used a so-called 7 bit latch circuit 7BL, a subtractor SUB and an up/down counter UDC interconnected as disclosed in Fig. 2.

The 7-bit latch circuit 7BL and the counter UDC are clocked (CL) together by the secondary frequency signal, which, however, is first passed through a flip-flop FF, which is clocked by the clock frequency CF, the latch circuit being clocked on each positive edge and being thereby latched to the current position of the frequency divider DC.

The aim of said flip-flop FF is to eliminate the hazard condition which may arise if said latch circuit or said up/down counter is clocked and the counter of the

31.8.1979

7

PHZ 78/012C

frequency divider is progressed one step simultaneously.

When the phase of the secondary frequency signal is changed the position of the latch circuit will be changed correspondingly. In order not to obtain an unstable indication due to noise and small disturbances, e.g. so that the position will vary between 0 and 127, said 7-bit up/down counter UDC has been introduced. The counter is then made to follow the latch circuit with some amount of a backlash, e.g. three units which corresponds to three of said subdistances and equals about $8 \times 0.03 = 0.24$ mm. This is obtained by comparing the positions of the counter and the latch circuit in a so-called "subber" (SUB), that it is an adder with one input signal inverted. If it follows from this comparison that the counter has a backlash which is more than e.g. seven units ($|DIFF| > 4$) the same will be advanced one step on the negative edge of the secondary frequency, or also the same is stepped backwards one step if the counter is leading by activating the counter via input EN and UPP/NED.

Thereby is obtained on the output of the up/down counter a noise free digital indication of the carriage position within a period of the Meander-structure. The zero crossings of this indication, that is the crossings between the periods of the Meander-structure, are obtained by sensing the carry bit CA of the 7-bit counter UDC. Then also the direction of the zero crossing is sensed, that is if up or down counting is in question.

Both of said output signals from the up/down counter are supplied to the microprocessor 12 as an indication of the carriage position.

In Fig. 3 is shown a signal diagram of signals in different points of the circuit diagram in Fig. 2.

Signal a shows the clock frequency signal CF having the frequency of 3.07 MHz.

Signal b shows the so-called primary frequency signal having the approximate frequency of 24 kHz, which is obtained by counting down signal a with a factor of 128 in the frequency divider DC. As shown in connection

31.8.1979

8

PHZ 78/012C

with signal a follows therefrom that each period of the b-signal comprises 128 periods of the a-signal. Each period of the a-signal may also be said to represent the sub-distance of 0.0302 mm defined above and being the smallest detectable change of the carriage position.

From the secondary element of the Meander-indicator the above mentioned signal e is obtained, said signal results after being pulse shaped in the converter C in the shown signal e'. The hazard of the edge position of e' is indicated by a number of different edge positions at each transition. In the case shown the carriage has moved one sub-distance = 0.0302 mm and therefore initiated a corresponding phase change of the second positive edge of said e'-signal.

Signal h from the counter in the frequency divider DC is a binary number of 7 bits assuming values between 0 and 127 in synchronism with signal a. The h-signal in Fig. 3 is represented by a pulse train of which each half period represents an indicated value of the h-signal. At the first and second shown positive edges of the e'-signal this h-signal takes the value of 2 and 3, respectively.

Signal g represents the e'-signal after having passed through the flip-flop FF. From the diagram is clear that this flip-flop is clocked on the negative edges of the a-signal.

Signal i represents the value of the binary number of seven bits comprised in the latch circuit 7BL. This signal is represented by a two-level signal having the signal value indicated at the transitions between said levels. From the diagram is clear that the latch circuit is clocked at the positive edges of signal g. Signals h and g then gives that the i-signal has the value of 2 between the shown first and second, respectively, edges of the g-signal and thereafter the values 3, 4, etc.

Signal j represents the value of the binary number appearing on the output of counter UDC by means of a two-level signal the values of which are indicated at the transitions.

31.8.1979

9

PHZ 78/012C

Signals i and j are compared in circuit SUB and if this comparison results in that counter UDC has a back lash to or leads the position of the latch circuit an amount which is greater than what corresponds to four units, i.e. four periods of signal a ($|DIFF| > 4$) the corresponding signal k of circuit SUB changes to a high level.

In the case shown the output signal j of counter UDC has initially the value of 126, while the i-signal has the value 2. Therefore the difference consists of the four values 127, 0, 1, 2, which is not greater than four units and consequently the k-signal will not change to its high level at the first appearing negative edge of g.

When later on the i-signal changes to the value 3 the requirement $|DIFF| > 4$ is fulfilled and the belonging output signal k changes to the high level.

In the diagram the assumption is that the carriage is moved in the one and same direction all the time and that appearing interferences are small, and so the difference between the i-signal and the j-signal will always be positive, which is indicated by a stationary high level of the signal l from circuit SUB, said signal being obtained from the output of the seventh bit thereof.

By the simultaneous high levels of signals l and k the requirement for advancement of UDC is fulfilled and this occurs at the next following negative edge of the g-signal, and thereat the j-signal will assume the value of 127 and at the same time k-signal will return to the low level.

Fig. 3 also illustrates what happens at a later stage when the carriage has been moved another 0.0302 mm and as a consequence thereof the output signal i of the latch circuit has obtained the value of 4. Since the j-signal has the value of 127 the requirement $|DIFF| > 4$ holds, and the k-signal assumes its high level. At the next following negative edge of the g-signal the UDC counter is advanced and signal l changes to the value of

31.8.1979

10

PHZ 78/012C

0 since a full period of the Meander-structure has been passed. At the same time a pulse is generated on the carry output of UDC which is indicated by the signal m.

5 The m-signal is supplied to the micro processor as an interrupt instruction initiating the micro processor to sense the signal j and to conclude therefrom the direction of the zero crossing by checking if the new position of UDC has the value of 0 or 127 and dependent thereon count up or down a stored value for the number of
10 zero crossings.

As described above an accurate value of the carriage position is accessible for the micro processor at each moment by signals j and m. In a matrix-printer this information is used for the initiation of relevant
15 printing needles.

Settling of the carriage position at the two occasions according to the invention is made in control of the micro processor program, and then the first occasion appears after the acceleration distance of the
20 carriage, which is known by the characteristic of the drive motor.

At the first occasion the carriage position is sensed as described above and at the same time a counter comprised in the micro processor is started, said
25 counter being thereafter advanced by the timing circuit TC thereof. In order to obtain a secure indication of the carriage position the position is sensed repeatedly and accepted first after reading of two equal values in succession. When the counter cycle has been run through
30 the second occasion occurs and the carriage position is again sensed in the same manner. The distance of movement is calculated as the difference between the recognized carriage positions and with a knowledge of the cycle time of the counter an actual value of the carriage speed is
35 calculated which is thereafter compared with a stored optimal value. Based on this comparison is thereafter calculated a correction factor of the reference signal Ref (Fig. 1) and a compensated Ref-signal is generated;

31.8.1979

11

PHZ 78/012C

said signal being supplied to the drive unit 10 via the D/A converter 11 during the next following carriage movement.

5 During the next following carriage movement a corresponding calculation of the actual carriage speed is made and dependent thereon is again calculated an eventually changed reference value. The operation is repeated in the described manner during each following carriage movement.

10 Above has been described a method and device for checking and eventually updating a carriage speed reference signal once per carriage movement. However, as should be obvious to anyone skilled in the art, the carriage speed may be sensed repeatedly during the carriage movement and corresponding reference signal values may be
15 calculated and supplied to the drive unit, provided that a high speed micro processor is available.

20

25

30

35

31.8.1979

1

PHZ 78/012C

CLAIMS:

1. A method for controlling the carriage movement in a printer comprising a printing head supported by the carriage (5) and which is moved along a record carrier by the carriage movement, in which the carriage is moved by
5 means of a direct current motor (6) supported by the movable carriage or, alternatively, being stationarily mounted in the printer, the direct current motor being supplied by a drive unit (10) which compensates for appearing variations of load during the carriage movement,
10 characterized in that the drive unit (10) is supplied with a reference signal (Ref) via a digital-to-analogue converter (11) from a microprocessor (12) comprised in the printer, the carriage being thereby moved along at a carriage speed which corresponds to said reference signal,
15 in that the position of the carriage along the record carrier is sensed at two occasions when the carriage has reached the constant carriage speed which corresponds to the reference signal, in that the position information, after being eventually digitalized, is supplied to the
20 microprocessor for calculating the actual carriage speed, in that the carriage speed so calculated is compared with an optimal carriage speed value which is stored in the microprocessor, and a compensation factor of the actual reference signal is calculated on basis of this comparison,
25 in that a reference signal which has been modified by said compensation factor is supplied to the drive unit during the next following carriage movement, and in that said compensation factor is checked at each carriage movement by repeating the indicated measures.
- 30 2. A device for controlling the carriage movement in a printer comprising a printing head supported by carriage (5), the printing head being moved along a record carrier by the carriage movement, a direct current motor (6)

31.8.1979

2

PHZ 78/012C

for moving the carriage which is supported by the movable carriage or, alternatively, being stationarily arranged in the printer, and a drive unit (10) for generating a drive signal to the direct current motor, in which the drive unit compensates for appearing load variations during the carriage movement, characterized in that the device further comprises a carriage position indicator (1, 2, 3) and a timing circuit (TC) for sensing the position of the carriage in relation to a position reference point at two occasions which are determined by said timing circuit, a signal processing circuit (4) for analogue-to-digital conversion of the carriage position values which are obtained and for the supply of said values to a microprocessor (12), said microprocessor being arranged so as to calculate therefrom an actual carriage speed between said occasions and to compare this speed and an optimal carriage speed, which is stored in the microprocessor, and to calculate on basis of this comparison, a compensation factor for a reference signal value which is generated by the micro processor and supplied to the drive unit (10) via a digital-to-analogue converter (11) during the next following carriage movement.

3. A device as claimed in Claim 2, characterized in that the carriage position indicator comprises a Meander-indicator comprising an element (1) which is fixedly mounted in the printer and being extended in the movement direction of the carriage (5) and having a length which corresponds to the printing width of the printer, and two further elements (2, 3) supported by the carriage and arranged close to and along the fixed element (1) in order to secure a good inductive coupling between the Meander loop structures of said elements, in which said two elements are provided on the carriage at a mutual distance which is such that the belonging loop structures will have a mutual displacement equal to $1/4$ period of the loop structure relative to the fixed element, and further comprising a sinus generator (SIN) and a cosinus generator (COS) for supplying each a respective one of the element

31.8.1979

3

PHZ 78/012C

on the carriage with sinus type signals of same frequency and having a mutual phase deviation of 90° , thereby generating inductively in said fixed element an output signal (e) of same frequency and having a phase deviation
5 (β) relative to the supplied sinus signal which is a measure of the carriage displacement within a period of the Meander-loop structure.

4. A device as claimed in Claim 3, characterized in that for the conversion of said phase deviation (β) to
10 a digital value the device comprises a counter (DC) having the cycle length of A bits, which is advanced at a frequency equal to A times the frequency of the sinus signal and in synchronism therewith, and a latch circuit (7BL) to which is supplied a binary number having A bits from
15 said counter and the value of which is consequently changed step-wise and periodically from 0 to the maximum value of the A bits number, and in that the sinus output signal from the fixed indicator element (1), after having been pulse-shaped to an rectangular wave with frequency and phase
20 unchanged, is supplied to said latch circuit as a clock signal, thereby locking the latch circuit synchronously with the output signal of the fixed indicator element to binary numbers providing at each occasion a binary representation of said phase deviation within a period of the Meander-loop
25 structure.

5. A device as claimed in Claim 4, characterized in that for the suppression of interferences in the carriage position indication supplied by said latch circuit (7BL) the device comprises a so-called "subber"-circuit,
30 which consists of an adder one input of which is inverted, and an up/down counter (UDC) having the cycle length of A bits and made to operate with a predetermined lag relative to the latch circuit, in which the output signals from the latch circuit and the up/down counter are supplied to the
35 +input and -input, respectively, of said "subber" circuit, comparing thereby said output signals, and in which, provided said comparison indicates that the lag is greater than the predetermined amount, the up/down counter is

0009291

31.8.1979

4

PHZ 78/012C

advanced one step in synchronism with the pulse-shaped
output signal from the fixed indicator element (1),
thereby allowing the up/down counter to provide an
undistorted, digital carriage position indication.

5

10

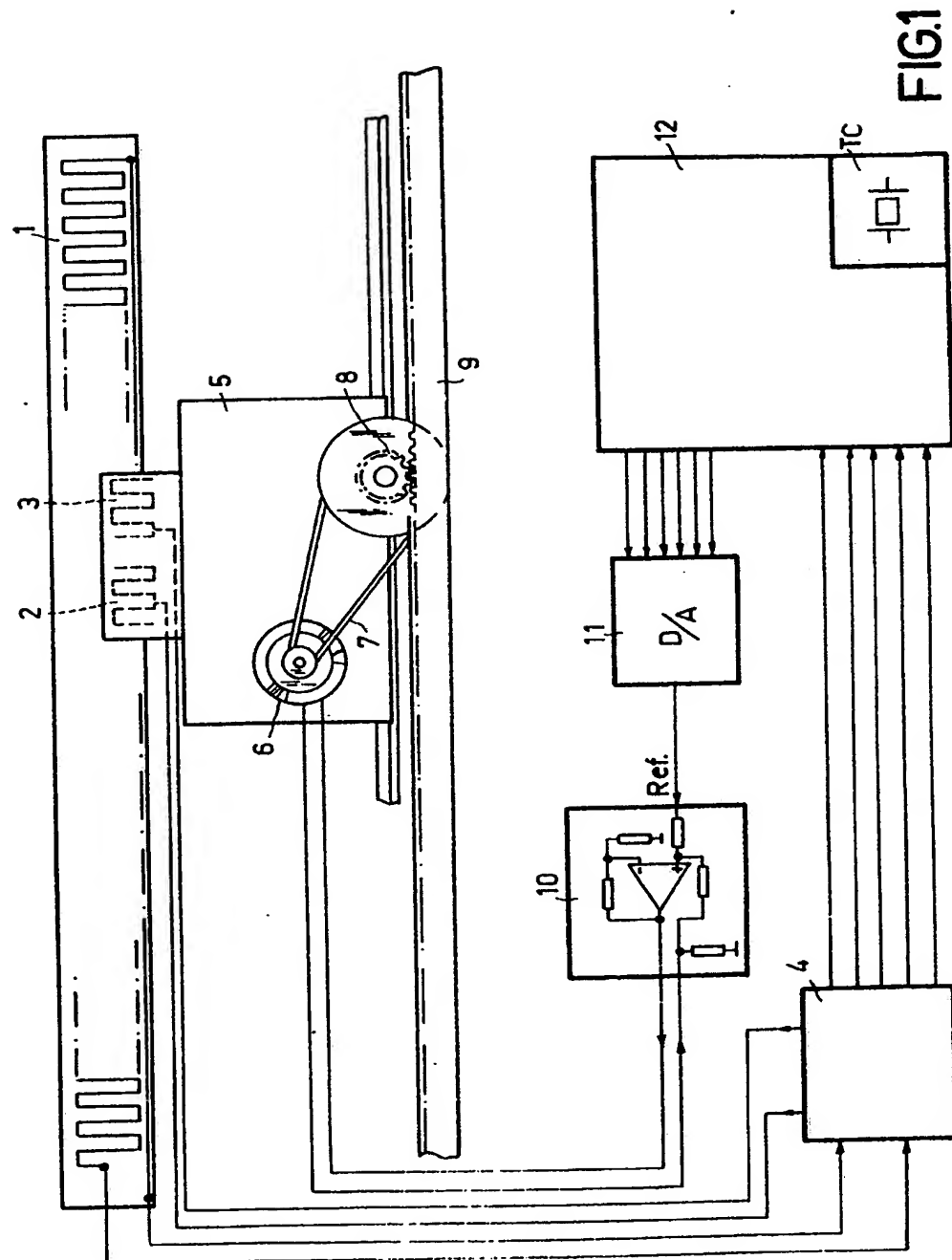
15

20

25

30

35



2/3

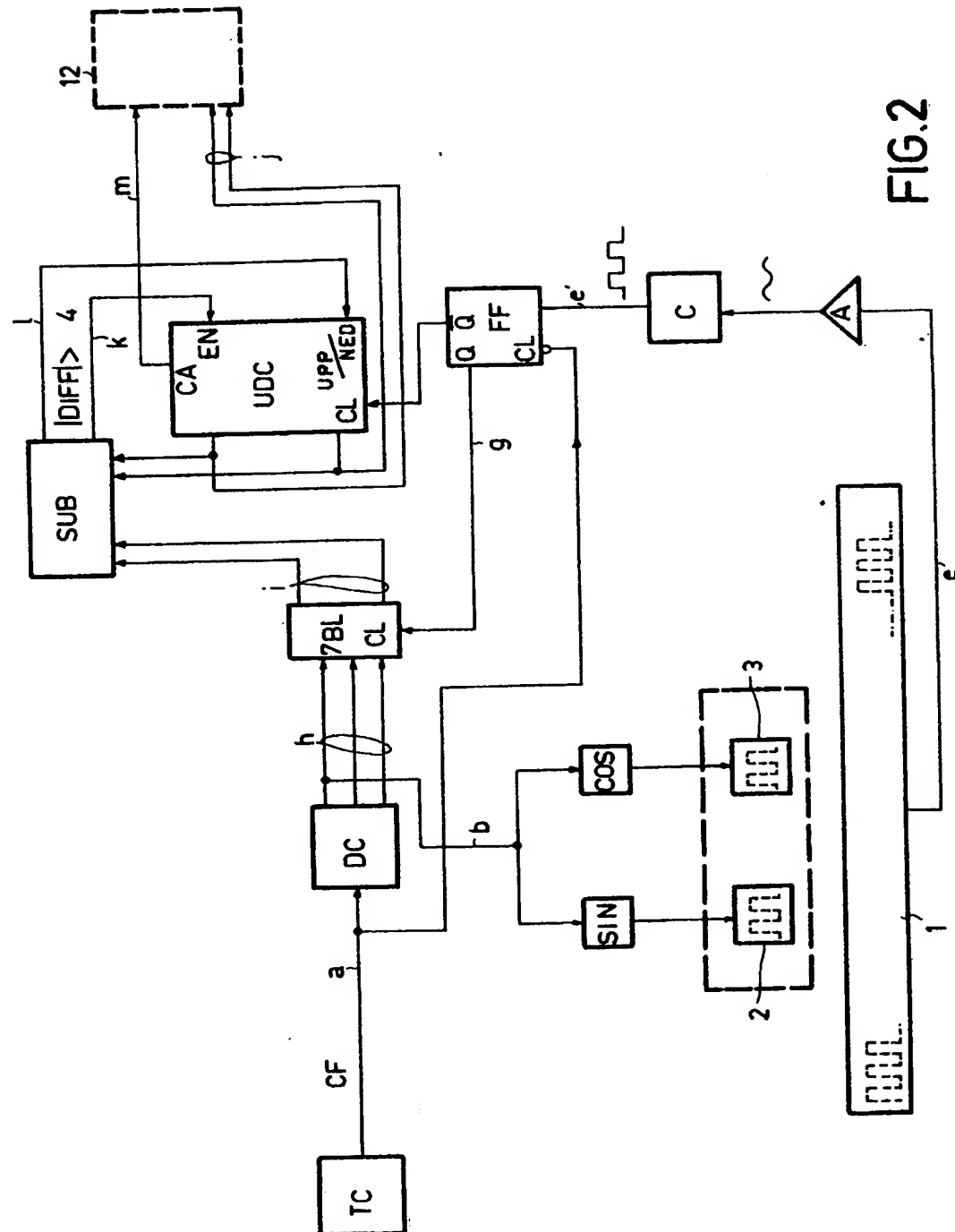
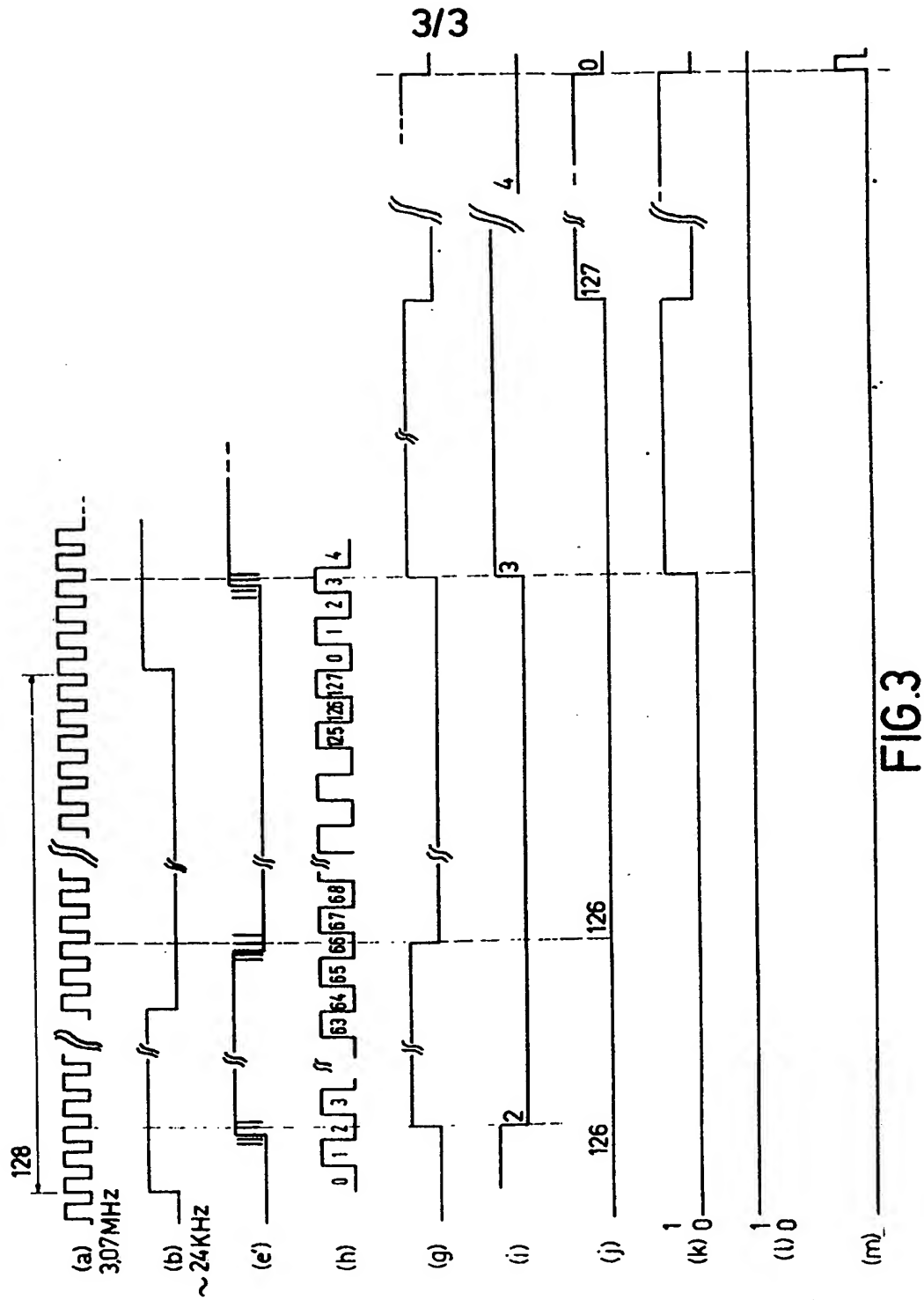


FIG. 2



0009291



European Patent
Office

EUROPEAN SEARCH REPORT

Application number

EP 79 200 520.9

DOCUMENTS CONSIDERED TO BE RELEVANT			CLASSIFICATION OF THE APPLICATION (Int. Cl.)
Category	Citation of document with indication, where appropriate, of relevant passages	Relevant to claim	
A	<u>US - A - 4 044 882</u> (WEINKE et al.) * complete document * --		G 06 K 15/02 B 41 J 19/30
A	<u>US - A - 3 942 619</u> (NORDSTROM et al.) * complete document * --		
A	<u>DD - A - 100 207</u> (SOCIETE D'APPLICATIONS GENERALES D'ELECTRICITE ET DE MECANIQUE) * complete document * --		TECHNICAL FIELDS SEARCHED (Int. Cl.)
A	<u>US - A - 3 539 895</u> (J.K. McGEE) * complete document * -----		B 41 J 19/00 B 41 J 21/00 B 41 J 23/00 B 41 J 25/00 G 06 K 15/02
			CATEGORY OF CITED DOCUMENTS X: particularly relevant A: technological background O: non-written disclosure P: intermediate document T: theory or principle underlying the invention E: conflicting application D: document cited in the application L: citation for other reasons
X The present search report has been drawn up for all claims			&: member of the same patent family, corresponding document
Place of search		Date of completion of the search	Examiner
Berlin		03-01-1980	ZOPF

**This Page is Inserted by IFW Indexing and Scanning
Operations and is not part of the Official Record**

BEST AVAILABLE IMAGES

Defective images within this document are accurate representations of the original documents submitted by the applicant.

Defects in the images include but are not limited to the items checked:

- ☐ BLACK BORDERS
- ☐ IMAGE CUT OFF AT TOP, BOTTOM OR SIDES
- ☐ FADED TEXT OR DRAWING
- ☒ BLURRED OR ILLEGIBLE TEXT OR DRAWING
- ☐ SKEWED/SLANTED IMAGES
- ☐ COLOR OR BLACK AND WHITE PHOTOGRAPHS
- ☐ GRAY SCALE DOCUMENTS
- ☐ LINES OR MARKS ON ORIGINAL DOCUMENT
- ☐ REFERENCE(S) OR EXHIBIT(S) SUBMITTED ARE POOR QUALITY
- ☐ OTHER: _____

IMAGES ARE BEST AVAILABLE COPY.

As rescanning these documents will not correct the image problems checked, please do not report these problems to the IFW Image Problem Mailbox.